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Capacity Enhancement for Irregular Reconfigurable Intelligent Surface- Aided Wireless Communications

Ruochen Su*, Linglong Dai*, Jingbo Tan*, Mo Hao#, and Richard MacKenzie+

***Department of Electronic Engineering, Tsinghua University**

#Tsinghua SEM Advanced ICT LAB, Tsinghua University

+BT Technology

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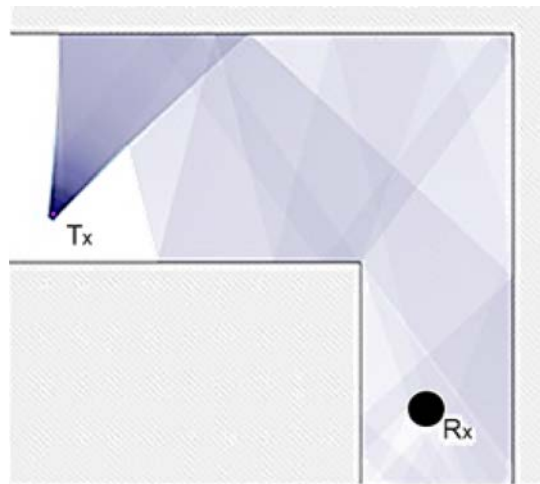
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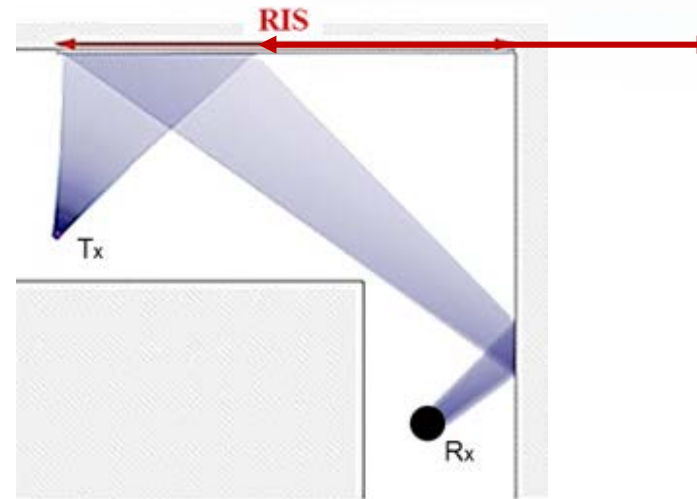
Background

- **Reconfigurable intelligent surface (RIS)**

- A two-dimensional electromagnetic **metasurface**
- **Control** the propagation of electromagnetic waves
- **Manipulate** the wireless environment to improve the quality of the signal



Traditional wireless communications:
Heavily rely on the environment



RIS-aided wireless communications:
Intelligently control the environment

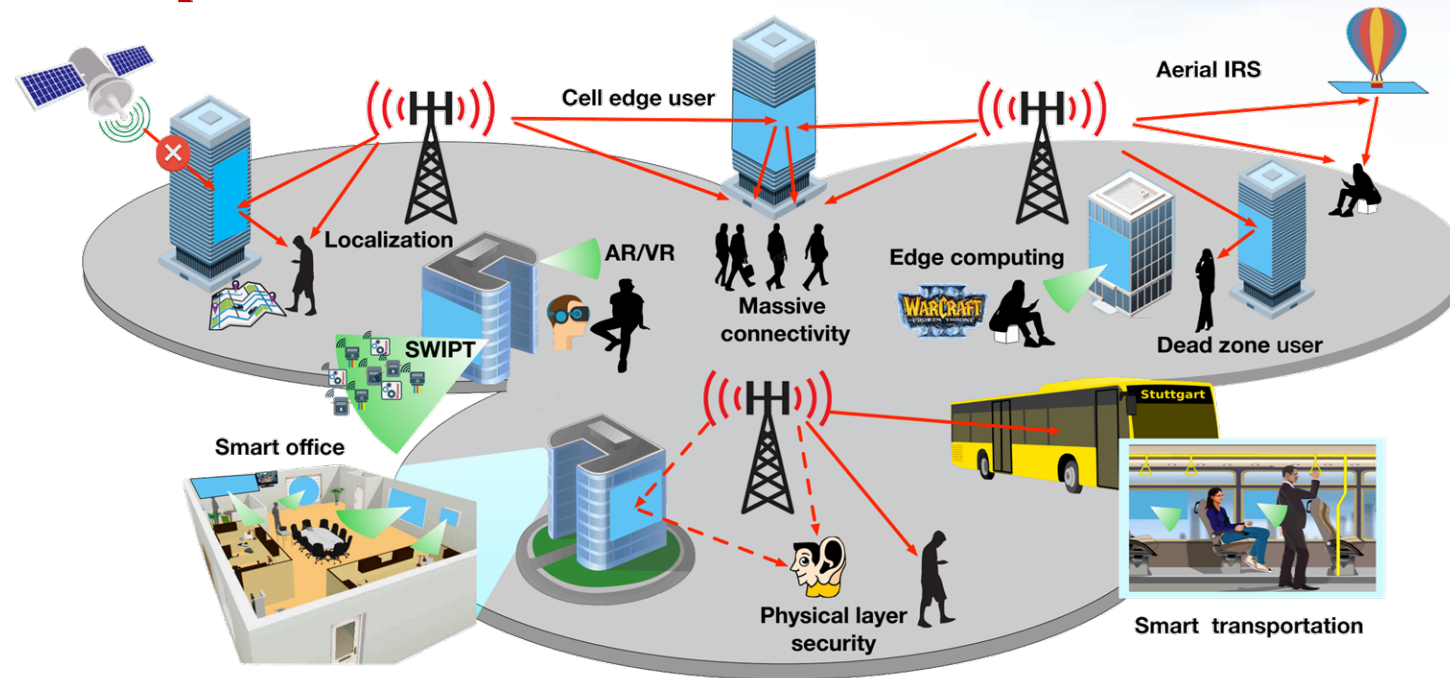
E. Basar, M. Di Renzo, J. De Rosny, M. Debbah, M. Alouini, and R. Zhang, “[Wireless communications through reconfigurable intelligent surfaces](#),” *IEEE Access*, vol. 7, pp. 116753-116773, Jul. 2019.



Background

● RIS-aided wireless communications

- Overcome the **blockage**
- Enhance the **signal quality**
- Save the **power consumption**



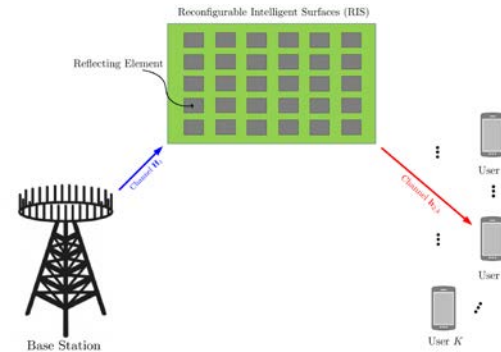
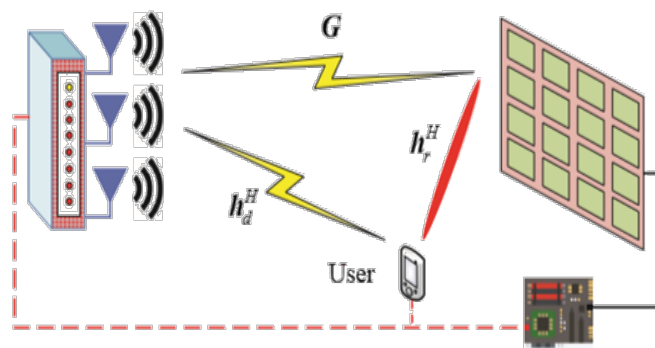
Q. Wu, S. Zhang, B. Zheng, C. You, R. Zhang, "Intelligent reflecting surface aided wireless communications: A tutorial," *arXiv preprint arXiv:2007.02759*, Jul. 2020.



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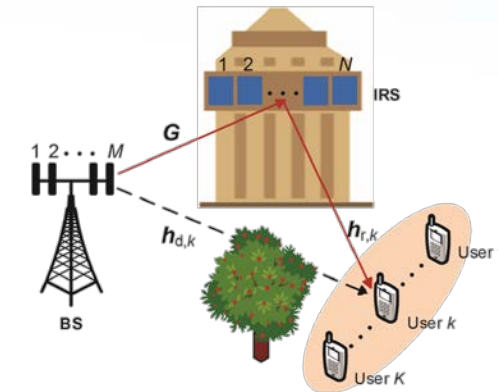
● Beamforming design

- Base station (BS) + RIS + user ends (UE)
- Line-of-sight path (BS-UE) + reflection path (BS-RIS-UE)
- Precoding (BS) + reflection coefficients (RIS)



● Optimization objective

- Sum-rate
- Energy efficiency
- Transmit power



[1] C. Huang, A. Zappone, G. C. Alexandropoulos, M. Debbah, and C. Yuen, “Reconfigurable intelligent surfaces for energy efficiency in wireless communication,” *IEEE Trans. Wireless Commun.*, vol. 18, no. 8, pp. 4157–4170, Aug. 2019.

[2] H. Guo, Y.-C. Liang, J. Chen, and E. G. Larsson, “Weighted sumrate optimization for intelligent reflecting surface enhanced wireless networks,” *arXiv preprint arXiv:1905.07920v2*, May 2019.

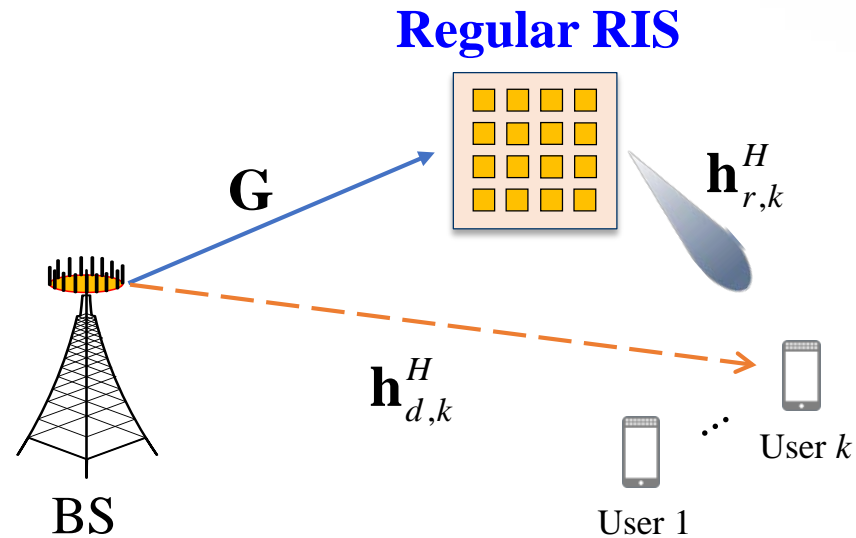
[3] Q. Wu and R. Zhang, “Beamforming optimization for wireless network aided by intelligent reflecting surface with discrete phase shifts,” *IEEE Trans. Commun.*, vol. 68, no. 3, pp. 1838–1851, Mar. 2020.



Background

● Challenge

- Prior works have only considered the regular RIS structure
- **Regular RIS**: High capacity requires a **large** number of RIS elements
- **Unbearable** system complexity and signal processing overhead



How to improve the **capacity** with a **limited** number of RIS elements?

[1] E. Basar, M. Di Renzo, J. De Rosny, M. Debbah, M. Alouini, and R. Zhang, "Wireless communications through reconfigurable intelligent surfaces," *IEEE Access*, vol. 7, pp. 116753-116773, Jul. 2019.

[2] C. Hu and L. Dai, "Two-timescale channel estimation for reconfigurable intelligent surface aided wireless communications," *arXiv preprint arXiv:1912.07990*, Dec. 2019.



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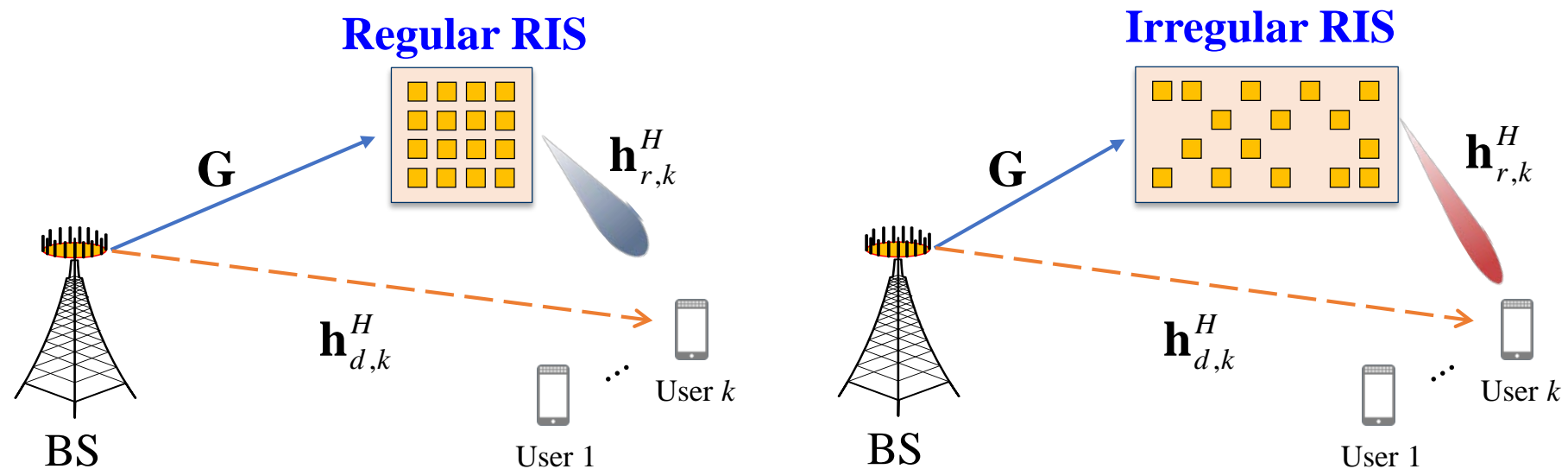
Capacity Enhancement by Irregular RIS

● Challenge

- **Regular array:** Elements are **regularly arranged** with constant interelement spacing
- High capacity requires a **large** number of RIS elements

● Proposal

- **Irregular array:** Elements are **irregularly arranged** on an enlarged surface
- Additional **degrees of freedom and spatial diversity** for more capacity



More **space** leads to more system **capacity**



System Model

● Irregular RIS-aided communications

- K **single-antenna users**, **BS** with M antennas
- **Irregular RIS** comprising N elements distributed over N_s grid points

$$\mathbf{y} = \left(\mathbf{H}_r^H \mathbf{Z} \Theta \mathbf{G} + \mathbf{H}_d^H \right) \mathbf{x} + \mathbf{n}, \quad \mathbf{x} = \sum_{k=1}^K \mathbf{w}_k s_k$$

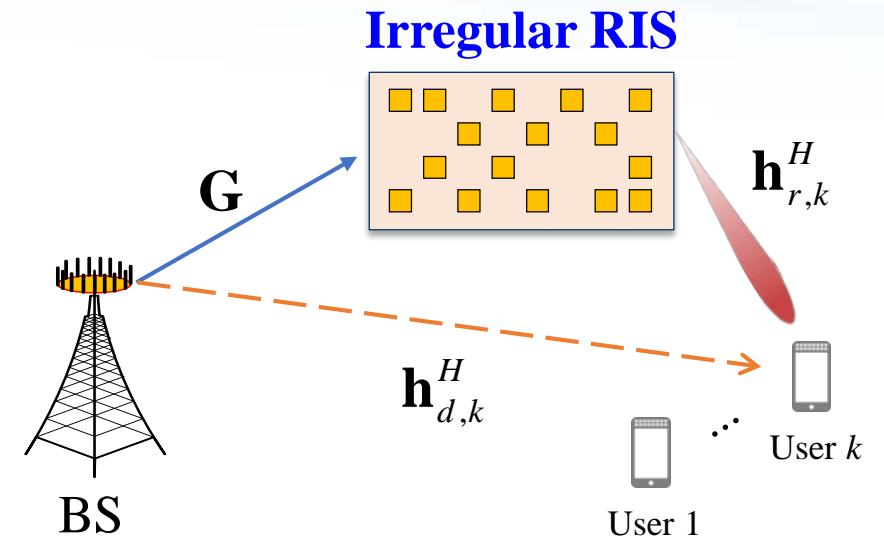
RIS topology

$$\mathbf{Z} = \text{diag}(\mathbf{z}), \quad \mathbf{z} = [z_1, z_2, \dots, z_{N_s}]^T, \quad z_i \in \{1, 0\}$$

$$\Theta = \text{diag}([\beta_1 e^{j\theta_1}, \beta_2 e^{j\theta_2}, \dots, \beta_{N_s} e^{j\theta_{N_s}}]), \quad \beta_n = 1, \quad \theta_n \in F = \{0, \pi\}$$

- The **signal-to-interference-plus-noise ratio (SINR)** of user k

$$\gamma_k = \frac{\left| \left(\mathbf{h}_{r,k}^H \mathbf{Z} \Theta \mathbf{G} + \mathbf{h}_{d,k}^H \right) \mathbf{w}_k \right|^2}{\sum_{i \neq k}^K \left| \left(\mathbf{h}_{r,k}^H \mathbf{Z} \Theta \mathbf{G} + \mathbf{h}_{d,k}^H \right) \mathbf{w}_i \right|^2 + \sigma^2}$$



System Model

● Channel model

$$\mathbf{y} = \left(\mathbf{H}_r^H \mathbf{Z} \mathbf{O} \mathbf{G} + \mathbf{H}_d^H \right) \mathbf{x} + \mathbf{n}, \quad \mathbf{H}_d^H = \left[\mathbf{h}_{d,1}, \mathbf{h}_{d,2}, \dots, \mathbf{h}_{d,K} \right]^H \in \mathbb{C}^{K \times M}, \quad \mathbf{H}_r^H = \left[\mathbf{h}_{r,1}, \mathbf{h}_{r,2}, \dots, \mathbf{h}_{r,K} \right]^H \in \mathbb{C}^{K \times N_s}$$

- **The small-scale fading:** uncorrelated Rayleigh fading channel model
- **The large-scale fading:** distance-dependent path loss
- **Path loss: BS-RIS-UE channel**

$$f_r(d_{\text{BR}}, d_{\text{RU}}) = C_r d_{\text{BR}}^{-\alpha_{\text{BR}}} d_{\text{RU}}^{-\alpha_{\text{RU}}},$$

Distance

Channel fading

- **Path loss: BS-UE channel**

$$f_d(d_{\text{BU}}) = C_d d_{\text{BU}}^{-\alpha_{\text{BU}}} \quad \text{Path loss exponent}$$

[1] O. Ozdogan, E. Björnson, and E. G. Larsson, "Intelligent reflecting surfaces: Physics, propagation, and pathloss modeling," *IEEE Wireless Commun. Lett.*, vol. 9, no. 5, pp. 581–585, May 2020.

[2] Z. Zhang and L. Dai, "A joint precoding framework for wideband reconfigurable intelligent surface-aided cell-free network," *arXiv preprint arXiv:2002.03744v2*, Feb. 2020.

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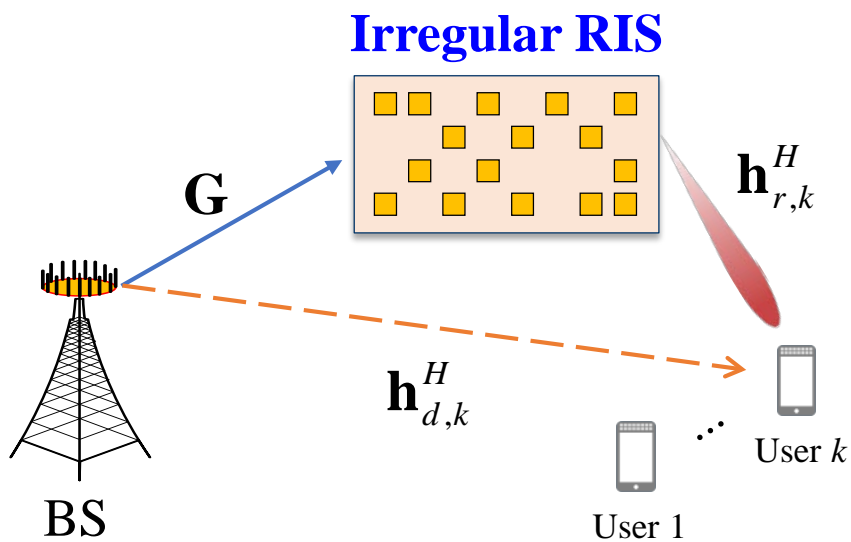
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Problem Formulation

- **Weighted sum-rate (WSR) maximization**

- **Joint optimization:** RIS topology and beamforming design
- The **topology design** lacks methodological guidance



$$\mathcal{P}_1 : \max_{\mathbf{z}, \mathbf{W}, \Theta} R = \sum_{k=1}^K \omega_k \log_2 (1 + \gamma_k)$$

Transmit power constraint

$$\text{s.t. } C_1 : \sum_{k=1}^K \|\mathbf{w}_k\|_2^2 \leq P_T,$$

Discrete phase shifts constraint

$$C_2 : \theta_n \in \mathcal{F}, \forall n = 1, 2, \dots, N_s,$$
$$C_3 : z_i(z_i - 1) = 0, \forall i = 1, 2, \dots, N_s,$$
$$C_4 : \mathbf{1}^T \mathbf{z} = N.$$

Non-convex: hard to solve

Sparsity constraints



Problem Formulation

● Solution

- **Decouple** the RIS topology design and the beamforming optimization
- For **a given topology**: Convert the original problem to P_2
- Let $\mathbf{Z}_0 = \mathbf{I}_N$: Equivalent to **regular RIS**-aided wireless communications

$$\mathcal{P}_2 : \begin{aligned} \max_{\mathbf{W}, \Theta} \quad & R = \sum_{k=1}^K \omega_k \log_2 (1 + \gamma_k) \\ \text{s.t.} \quad & C_1 : \sum_{k=1}^K \|\mathbf{w}_k\|_2^2 \leq P_T, \\ & C_2 : \theta_n \in \mathcal{F}, \forall n = 1, 2, \dots, N_s, \\ & C_3 : \mathbf{Z} = \mathbf{Z}_0. \end{aligned}$$

Given RIS topology

$$\gamma_k = \frac{\left| \left(\mathbf{h}_{r,k}^H \mathbf{Z} \Theta \mathbf{G} + \mathbf{h}_{d,k}^H \right) \mathbf{w}_k \right|^2}{\sum_{i \neq k} \left| \left(\mathbf{h}_{r,k}^H \mathbf{Z} \Theta \mathbf{G} + \mathbf{h}_{d,k}^H \right) \mathbf{w}_i \right|^2 + \sigma^2}$$

The SINR of user k

Precoding (BS)

Reflection coefficients (RIS)



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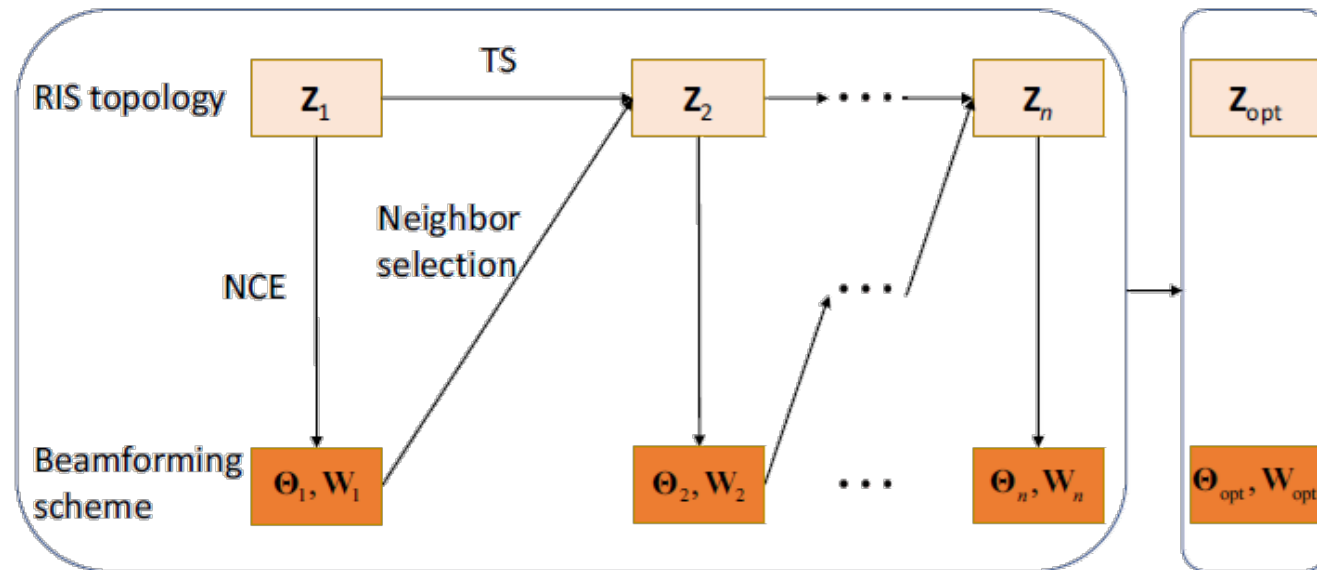
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Joint Optimization Framework

● Overview

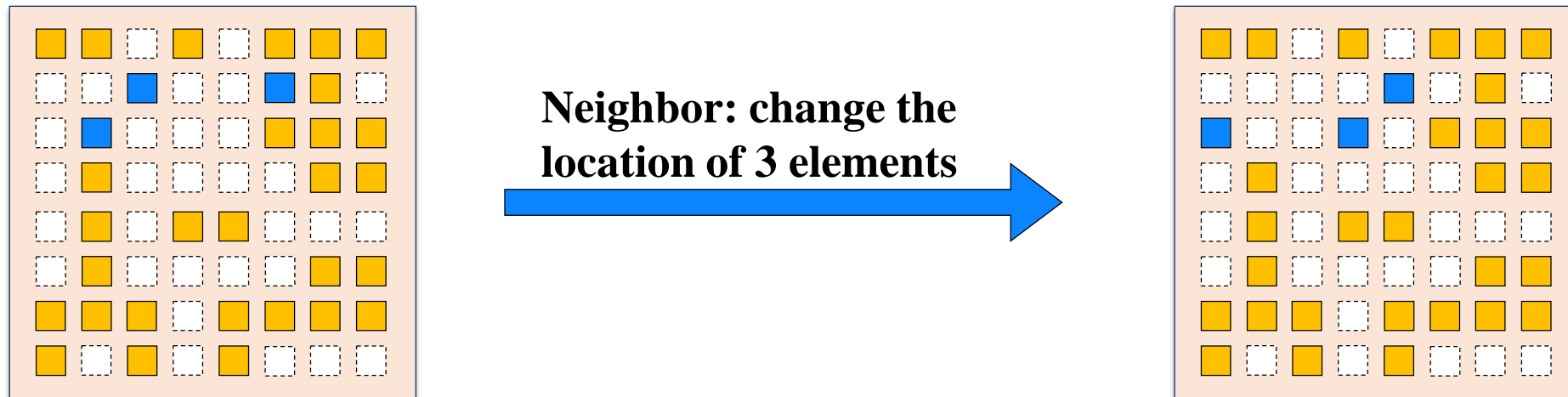
- **Alternating optimization**: Decouple the decision variables in P_1
- **RIS topology**: Tabu search (TS) method
- **Beamforming**: Neighbor extraction-based cross-entropy (NCE) method
- **General solution** to the classical sum-rate optimization problem



Joint Optimization Framework: RIS Topology

● TS-based sparse deployment of RIS

- **Input:** Tabu list, storage size, neighbor distance, neighborhood size, iterations
- **Output:** Optimal RIS topology Z
- Generate **alternative neighbors**
- Select the candidate with the maximum WSR in each iteration



X. Gao, L. Dai, C. Yuen, and Z. Wang, "Turbo-like beamforming based on tabu search algorithm for millimeter-wave massive MIMO systems," *IEEE Trans. Veh. Technol.*, vol. 65, no. 7, pp. 5731–5737, Jul. 2016.

Joint Optimization Framework: Beamforming Scheme

● NCE-based beamforming optimization

- **Input:** RIS topology, iterations, number of candidates/elites, quantized phase shifts set
- **Output:** Phase shifts matrix Θ , precoding matrix \mathbf{W}
- Generate **candidates** based on the probability distribution function

$$\Xi(\Theta; \mathbf{P}^{(i)}) = \prod_{n=1}^{N_s} \left(\prod_{k=1}^{2^b} (p_{n,k}^{(i)})^{\delta(\theta_n - F(k))} \right)$$

- **Neighbor extraction:** Change each effective element of the current optimal Θ in each iteration
- **Weighted probability transfer criterion**

$$\mathbf{P}^{(i+1)} = \arg \max_{\mathbf{P}^{(i)}} \frac{1}{C_{\text{elite}}} \sum_{c=1}^{C_{\text{elite}}} \eta_c \ln \Xi(\Theta^{(c)}; \mathbf{P}^{(i)}), \quad \eta_c = \frac{R(\Theta^{(c)}) C_{\text{elite}}}{\sum_{c=1}^{C_{\text{elite}}} R(\Theta^{(c)})}$$

● BS: Zero forcing precoding

$$\mathbf{W} = \mathbf{H}_{\text{eq}}^H (\mathbf{H}_{\text{eq}} \mathbf{H}_{\text{eq}}^H)^{-1} \mathbf{P}_B \frac{1}{2}$$

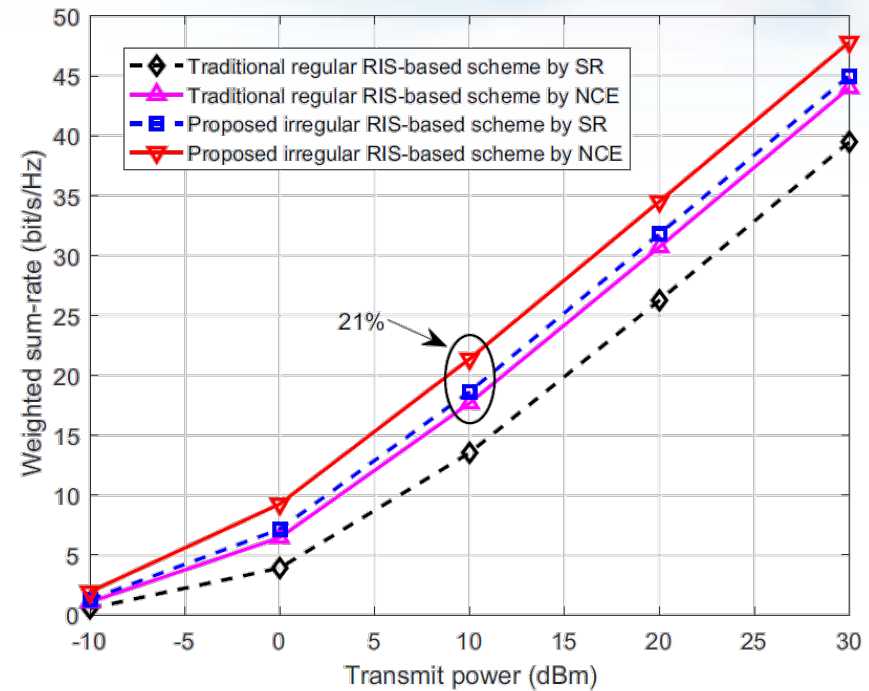
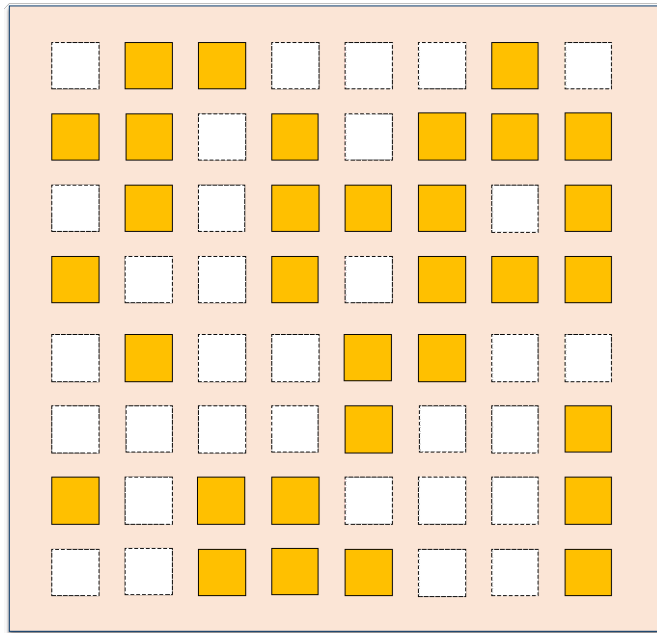
R. Y. Rubinstein and D. P. Kroese, *The cross-entropy method: A unified approach to combinatorial optimization, Monte-Carlo simulation and machine learning*, Springer Science and Business Media, 2013.



Simulations

- **WSR versus the transmit power**

- Antennas of BS: $M=4$, users: $K=4$
- RIS elements: $N=32$, grid points: $N_s=64$



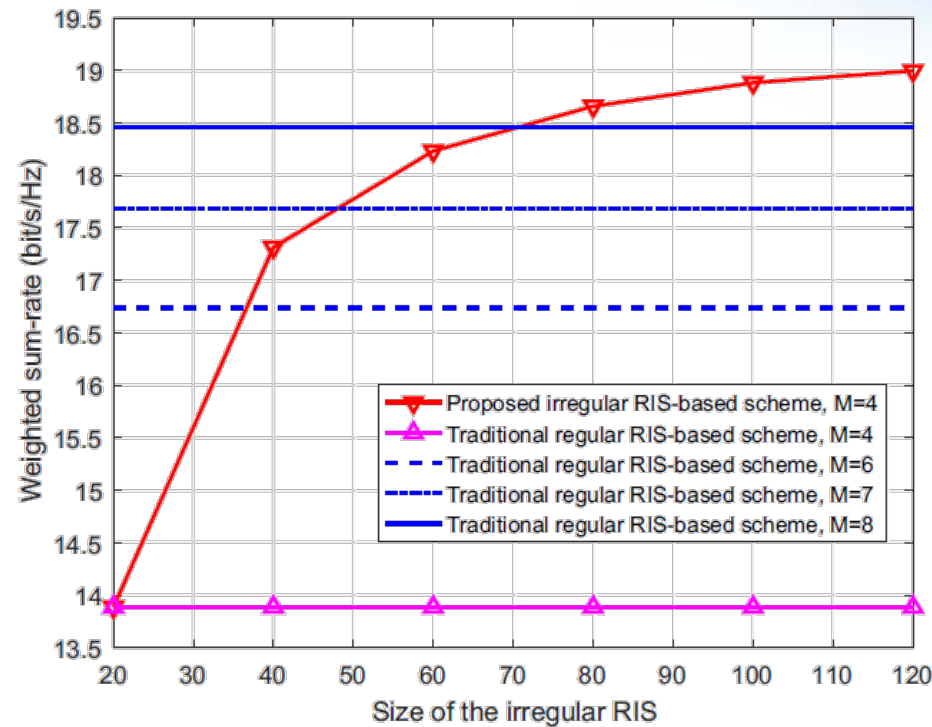
The **irregular** RIS outperforms the **regular** RIS



Simulations

● WSR versus the size of the irregular RIS

- Antennas of BS: $M=4$, users: $K=4$
- RIS elements: $N=32$



The **sparse ratio** of RIS: **tradeoff** between the cost and the performance



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Summary

● Challenge

- **Regular RIS**: High capacity requires a large number of RIS elements
- **Unbearable** system complexity and signal processing overhead

● Solution

- **Irregular RIS**: Additional degrees of freedom in **space** for more capacity
- Propose a joint optimization framework

● Conclusion

- The proposed irregular RIS can significantly **improve the system capacity** compared to the traditional regular RIS

● Further works

- The influence of mutual coupling effect at the RIS
- Channel estimation for irregular RIS-aided communications



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Thank you

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Ruochen Su

Department of Electronic Engineering

Tsinghua University

Email: src18@mails.tsinghua.edu.cn

Website: oa.ee.tsinghua.edu.cn/dailinglong

