A Tripolarization Antenna Fed by Proximity Coupling and Probe

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Abstract—The design and characterization of a conformal tripolarization antenna with three independent ports and three orthogonal polarizations is presented. The antenna uses both proximity (slot) coupling and probe-fed ports to reduce isolation and improve polarization purity. The measured bandwidth of the $-10$-dB return loss is found to be 190 MHz ($2.4$–$2.59$ GHz). Measurements across the working bandwidth also show that the isolations among the three ports are better than $-16$, $-30$, and $-40$ dB, respectively. Design parameters and experimental results are presented.

Index Terms—Probe, proximity coupling, tripolarization antenna.

I. INTRODUCTION

There are normally two independent polarizations associated with uniform plane waves. However, in a multipath propagation environment, which is the default favorable condition of multiple-input–multiple-output (MIMO) systems, a specific location may be expected to be illuminated by plane waves from multiple directions. In such circumstances, there will be three independent polarizations, and hence, a tripolarization antenna will need to be used to receive all available information and take full advantage of the potential of a MIMO system in increasing channel capacity. Recently, there has been increasing interest in the design and characterization of multipolarization antennas. For example, a tripolarization antenna composed of a dual-polarization circular patch and a monopole wire antenna was adopted by Das et al. [1]. A tripolarization antenna constructed from two orthogonal slots and a monopole wire was reported by Itoh [2]. A tripolarization antenna formed by a dual-polarization dielectric resonator and a monopole wire was also introduced by Gray [3]. In these cases, a quarter-wavelength monopole antenna had been used for the vertically polarized element [1]–[3]. To help design a tripolarized antenna with a low profile that would be easier to integrate into commercial products (e.g., cell phones), Zhong proposed a new probe-feed conformal tripolarization antenna [4]. The port isolations of this conformal antenna [4], however, were unsatisfactory, and additional design changes were needed to improve the performance. Proximity coupling, which was used by Steven [5] to design a dual-polarization patch antenna, is adopted in the new design presented in this letter to improve the isolation in a conformal tripolarization antenna. The measured isolations among three ports of the proposed antenna are found to be better than $-16$, $-30$, and $-40$ dB, respectively. The total height of the antenna is $10.8$ mm, and the area of the antenna is $94 \times 94$ mm$^2$.

II. CONFIGURATION OF THE TRIPOLARIZATION ANTENNA

Fig. 1 shows the configuration of the tripolarization antenna. The substrate consists of two dielectric substrate layers...
A dual-polarized circular microstrip patch sits on the top side of substrate 1. Two H-shaped slots, which provide the proximity coupling between feed lines and patches, are etched in the ground plane between substrates 1 and 2. Two microstrip feed lines are placed on the bottom side of substrate 2 below each slot for producing two orthogonally polarized radiation patterns. The two H-shaped slots are arranged in a “T” shape in order to improve the isolation between two polarizations. These two ports are marked as P1 and P2, respectively. In the center of the patch antenna, there is a grounded hole through two dielectric substrate layers orthogonally to the patch. A coaxial line passes through the hole and feeds a disk (capacitive)-loaded monopole antenna on the top of the patch. The inner conductor of the coaxial line is connected with the disk-loaded monopole antenna. The patch antenna acts as a ground for the monopole antenna. The monopole port is marked as M1. The height of the monopole antenna is 5 mm. The diameter of the monopole circular patch is 15.6 mm. With the height of the monopole antenna getting lower, the shunt capacitance between the disk and the patch antenna becomes higher. In order to improve impedance matching at port M1, a 1.5-nH shunt inductor is selected by experiment as a matching component to counteract the capacitance.

The dielectric constant is 2.6 for substrate 1 and 4.5 for substrate 2. The width of microstrip lines that feed P1 and P2 is 1.5 mm and designed as 50 Ω. The dimensions of two dielectric substrate layers are defined by the parameters $W$, $L$, $R$, $h_1$, and $h_2$. The dimensions of the “H” slot are defined by $d_s$, $l_s$, $l_{a1}$, $l_{a2}$, $w_{a1}$, $w_{a2}$, and $D$. The size of the monopole antenna is defined by $h_0$ and $r$. The prototype antenna designed has the following parameters: $W = L = 94$ mm, $R = 40$ mm, $h_1 =$
5 mm, \( h_2 = 0.8 \) mm, \( h_0 = 5 \) mm, and \( r = 15.6 \) mm. Port 1 has
the following parameters: \( d_{s1} = 32.5 \) mm, \( l_s = 7 \) mm, \( l_{a1} = 17 \) mm, \( l_{a2} = 2 \) mm, \( w_{a1} = 1 \) mm, and \( w_{a2} = 4 \) mm. Port 2
has the following parameters: \( D = 12.3 \) mm, \( d_{s2} = 47 \) mm,
\( l_s = 8.8 \) mm, \( l_{a1} = 11.4 \) mm, \( l_{a2} = 2 \) mm, \( w_{a1} = 1 \) mm, and
\( w_{a2} = 4 \) mm.

The E-fields of the two orthogonal patch modes are parallel to
the ground, and the monopole antenna radiates an E-field that is
perpendicular to the ground plane. Thus, this antenna provides
three orthogonal radiation modes.

### III. RESULTS

To verify the design, a prototype antenna is manufactured
and measured. Fig. 2 shows a photograph of the tri-polarization
antenna. The measured return losses for all three ports are
shown in Fig. 3(a). The \(-10\)-dB bandwidth of \( S_{11} \) (port P1),
\( S_{22} \) (port P2), and \( S_{33} \) (port M1) are 2.38–2.62, 2.40–2.61, and
2.38–2.59 GHz, respectively. The isolations between any two
ports are shown in Fig. 3(b). The isolation between ports P1
and M1 is below \(-16\) dB in the operational bandwidth. The isolations
between ports P1 and P2 or P2 and M1 are better than
\(-30\) and \(-40\) dB, respectively. These data represent significant
improvements over the results given in the earlier attempt
described in [4]. The much higher coupling between P1 and M1
than between P2 and M1 is due to the asymmetric layout of the
feeding structure. If using the feeding pin of port M1 as a rea-
ference, all E-fields inside the slot of port P1 are along the same
direction, thus generating higher coupling. All E-fields inside the
slot of port P2 are perpendicular to the feeding pin. The fields
on both sides of the symmetric line of the longer dimension are
canceling each other, thus producing better isolation.

Fig. 4 shows the measured radiation patterns for each port. It
clearly shows that the electric field excited by port P1 is parallel
to the \( X \)-axis, the E-field excited by port P2 is parallel to the
\( Y \)-axis, and the E-field excited by port M1 is parallel to the
\( Z \)-axis. The designed antenna, therefore, has three orthogonally
polarized modes.

### IV. CONCLUSION

This letter describes the design and characterization of a com-
 pact tri-polarization antenna whose total height is 10.8 mm. The
isolation is below \(-16\) dB between port P1 and port M2, below
\(-30\) dB between port P1 and P2, and below \(-40\) dB between
port P2 and M1. The operational bandwidth is 190 MHz.

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May 2003.
